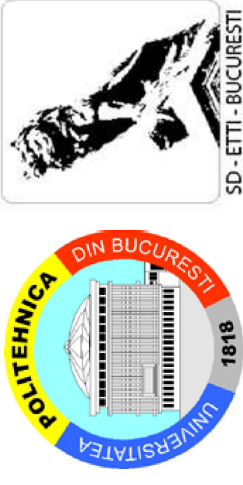


ANTENNA GAIN EVALUATION BASED ON WEIGHTING NEAR-FIELD MEASUREMENTS

Liliana Anchidin^(1,2), Razvan D. Tamas^(1,2), Adrian Androne⁽²⁾, and George Caruntu⁽¹⁾



(1) Department of Electronics and Telecommunications,
Constanta Maritime University,
C.P. 202, O.P. 2, 900780 Constanta, Romania



(2) Doctoral School of Electronics and Telecommunications,
Politehnica University of Bucharest,
Bd. Iuliu Maniu 1-3, Sect. 6, 061071, Bucharest, Romania

Email: liliana.achitei@cmu-edu.eu, razvan.tamas@cmu-edu.eu, tamas@ieee.org

Abstract — Antenna gain is generally extracted from near-field measurements when the size of the measuring site (anechoic chamber or open area test site) does not meet the far-field constraints that include both distance and probe size limitation. Near-field measurements are usually performed by scanning the field on a closed surface around the antenna with a small size probe. In this paper, we show that the distance averaging method that we have previously proposed for gain evaluation in a multipath environment can also be employed for measurements in the near-field zone. We introduce an alternative technique to extract the gain by applying weighting functions on the near-field data when the size of the probe cannot be neglected compared to the wavelength. We validated our method by comparing the results to the gain measured at distances in the far-field region.

Keywords— *Antenna gain; near-field measurements; weighting functions; averaging method*

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Liliana Anchidin was born in Dragoiesti, Romania, on September 29, 1987. She is a Ph.D. student with the Doctoral School of Electronics and Telecommunications at *Politehnica* University of Bucharest, Romania. She is currently an assistant professor with the Electronics and Telecommunications Department at Constanta Maritime University. Her field of interest includes antenna theory, more specifically near-field characterization techniques. Liliana Anchidin is a Graduate Student Member of the IEEE.



Razvan D. Tamas was born in Navodari, Romania, on May 10, 1970. He received the Ph.D. degree jointly from Politehnica University of Bucharest, Romania, and from Institut Polytechnique de Grenoble, France, and the Habilitation degree (Habilitation à Diriger des Recherches) from Institut Polytechnique de Grenoble, France. He is currently a Professor, head of the Electronics and Telecommunications Department at Constanta Maritime University, Romania, and a Ph.D. supervisor with the Doctoral School of Electronics and Telecommunications at Politehnica University of Bucharest, Romania. He was formerly a Senior Researcher with the Navy Scientific Research Center of the Military Equipment and Technology Research Agency, Romania. He also worked with the Institute of Microelectronics, Electromagnetism and Photonics of Grenoble, France on the R.N.R.T. AUBADE project and he was an Invited Professor with Université de Rennes 1 and Université Joseph Fourier, France. His field of interest includes antenna theory, numerical methods for electromagnetism, time-domain characterization, electromagnetic compatibility, and naval communication systems. Dr. Razvan D. Tamas is a Senior Member of the IEEE. He founded the IEEE AP-S Chapter in Romania and he currently serves as a Chair.

Adrian Androne is a Ph.D. student with the Doctoral School of Electronics and Telecommunications at *Politehnica* University of Bucharest, Romania. His field of interest includes antenna theory, more specifically frequency-selective surfaces.

George Caruntu was born in Constanta, Romania, on July 6, 1949. He received the Ph.D. degree from Politehnica University of Bucharest, Romania. He is currently a Professor with the Electronics and Telecommunications Department at Constanta Maritime University, Romania. His field of interest includes transducers and measurement techniques. Dr. George Caruntu is a Member of the IEEE.

I. INTRODUCTION

Gain measurements are usually performed by placing a calibrated probe as far from the antenna under test, as far-field constraints would be fulfilled.

Compliance with that limitation might not be possible when large antennas are measured.

Most of the measuring methods [1], [2], [3] use small probes placed on a surface surrounding the antenna under test, and near-field to far-field transforms are applied on the measured data.

We have previously presented [4], [5], [6] a method for antenna gain evaluation in a multipath environment, based on averaging data measured at different distances between the probe antenna and the antenna under test. Weighting on measured data was solely used to compensate the effects of the propagation (i.e., delay and attenuation), as it would be in the far-field region.

In this paper, we show that distance averaging can successfully be applied on a data set entirely acquired in the near-field zone with a probe of a size comparable to the wavelength. An alternative technique for processing near-field data, based on weighting functions, is proposed.

We firstly set up a far-field limit for transmission between two antennas of finite size.

Then we define the weighting functions to be applied on the measured, near-field data in order to remove the effect of the phase deviation between different pairs of source points and field points.

The method was validated by comparing weighted data measured in the near-field region to data measured on a set of distances mostly in the far-field region.

II. NEAR-FIELD TRANSMISSION BETWEEN TWO FINITE SIZE ANTENNAS

Let us consider the transmission between two linear dipoles of a total length $2h_1$ and $2h_2$, respectively (Fig. 1).

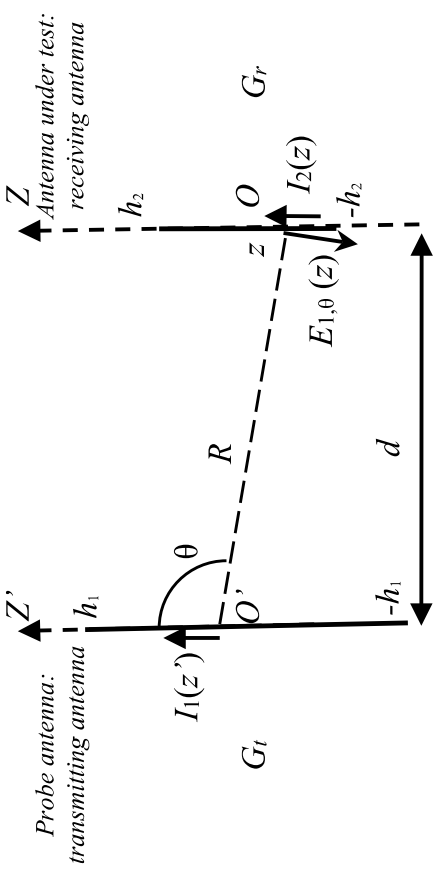


Fig. 1 Near-field transmission between two linear dipoles

II.1 Revision of the far-field limit for transmission between two antennas of finite sizes

The maximal phase deviation between two waves incident on the receiving antenna occurs when the field point is located on its top i.e., $z = h_2$, and the source points are located one at the same height as the field point and the other one at the bottom end of the transmitting antenna, respectively; that is, $z_1' = h_2$ and $z_2' = -h_1$.

By assuming that $d \geq 3(h_1 + h_2)$ the path length difference between the two incident waves can be expressed as:

$$R_{\max} - d \simeq \frac{(h_1 + h_2)^2}{2d}.$$

As for a single antenna the maximal phase deviation can be set at $\pi/8$ [7] as a reasonable error margin. It comes out that the far-field range for the configuration given in Fig. 1 is

$$d \geq \frac{8(h_1 + h_2)^2}{\lambda}$$

II.2 Definition of weighting functions for near-field data

We define weighting functions that near-field measurements should be multiplied by in order to assess the far-field. For a given direction one can define on both receiving and transmitting antennas constant, equivalent current distributions along the main axis of polarization.

The weighting functions can be derived by expressing the normalized voltage at the output of the receiving antenna from the mutual impedance [8], as follows

$$V_{0, norm} = d \exp(jk_0 d) V_0 \propto d \int_{-h_1}^{h_1} \int_{-h_2}^{h_2} I_1(z') I_2(z) \sin^2 \theta \frac{\exp[-jk_0(R-d)]}{R} dz dz' = d^3 \int_{-h_1}^{h_1} \int_{-h_2}^{h_2} I_1(z') I_2(z) \frac{\exp[-jk_0(R-d)]}{R^3} dz dz',$$

$$V_{0, norm}^\infty = \lim_{d \rightarrow \infty} d \exp(jk_0 d) V_0 \propto \int_{-h_1}^{h_1} \int_{-h_2}^{h_2} I_1(z') I_2(z) dz dz',$$

$$F(f, d) = \frac{V_{0, norm}^\infty}{V_{0, norm}} = \frac{4h_1 h_2}{d^3 \int_{-h_1}^{h_1} \int_{-h_2}^{h_2} \frac{\exp[-jk_0(R-d)]}{R^3} dz dz'}.$$

II.3 Definition of the average transfer function and AUT gain extraction

An average transfer function can be calculated from the scattering parameters measured at different distances between antennas:

$$|\overline{S_{21}}| = \sqrt{\frac{1}{N} \sum_n^N \left[\frac{d_n}{d_0} \cdot |S_{21,n}| \cdot |F(f, d_n)| \right]^2}$$

where d_n is a set of N distances, d_0 is the reference distance (usually set at 1 m), and $S_{21,n}$ are the corresponding transfer functions.

The AUT gain can then be extracted as

$$G_r = \frac{1}{G_t} \left(\frac{4\pi d_0}{\lambda} \right)^2 \frac{R_0}{R_{a2}} \frac{|\overline{S_{21}}|^2}{|1 - S_{22}|^2 (1 - |S_{11}|^2)}$$

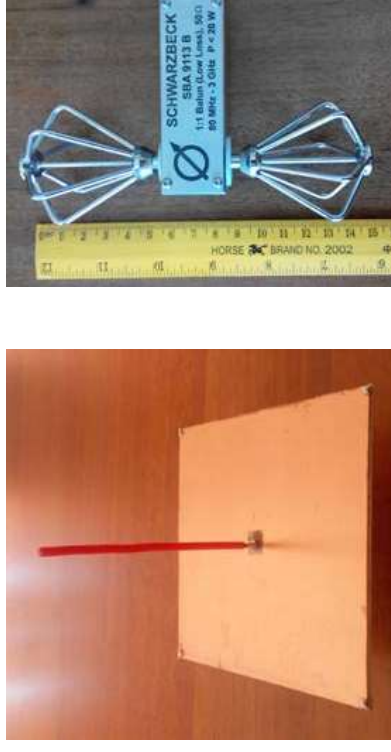
with R_0 the normalizing impedance and R_{a2} the real part of the input impedance of the antenna under test.

III. EXPERIMENTAL RESULTS

We consider a typical setup consisting of an antenna under test, a probe antenna, and a vector network analyzer. Measurements were performed in a reflection free environment in order to only investigate the impact of the near-field zone.

As an antenna under test, we took a monopole on a square ground plane. The monopole was 8.4 cm high and the side of the ground was 10 cm long (Fig. 2a). The antenna under test resonates around 800 MHz and 2.4 GHz. A calibrated, biconical dipole was employed as a probe antenna (Fig. 2b). The total height of the biconical dipole was $2h_1=14$ cm.

By taking into account a full size image for the monopole under test the far-field region at 1GHz starts at 68 cm between antennas.



a) b)

Fig. 2 Experimental setup: antenna under test (a), and probe antenna (b)

We performed two sets of measurements: one for distances between antennas ranging from 100 to 140 cm, and the other one for distances comprised between 15 and 60 cm. The distances in the first set are all in the far-field zone for frequencies up to 1.6 GHz; the second measuring range is completely below the far-field limit at 1 GHz and above.

The weighting functions for both sets are shown in Fig. 3 and Fig. 4, respectively.

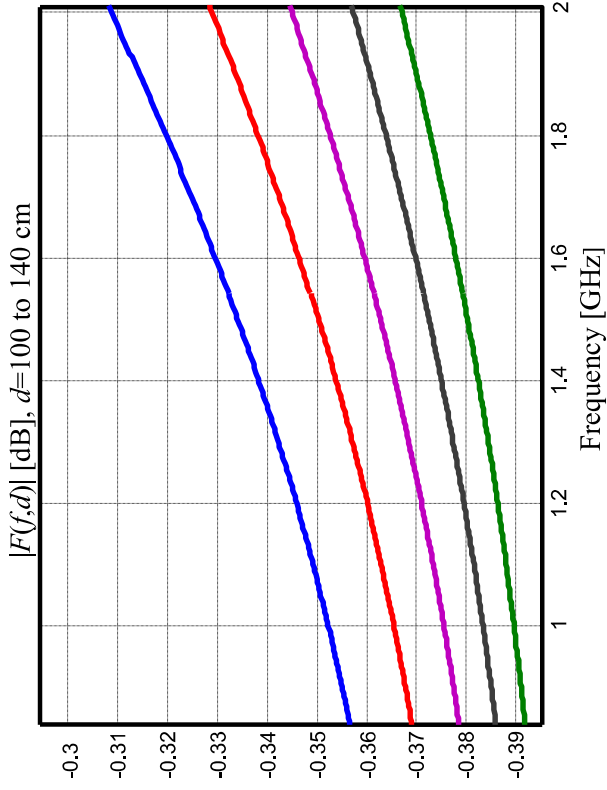


Fig. 3 Weighting functions for $d=100$ to 140 cm

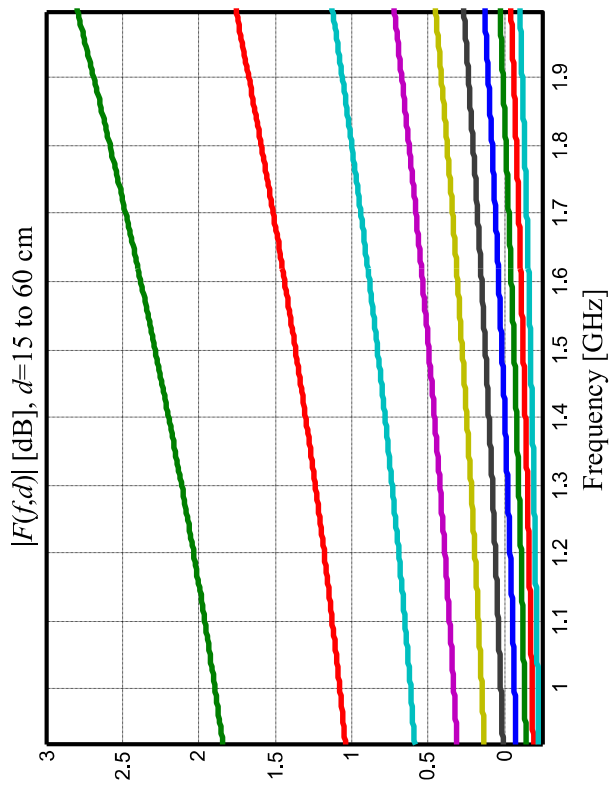


Fig. 4 Weighting functions for $d=15$ to 60 cm

For distances in the first set (100 to 140 cm) and for a frequency range of 0.85 to 2 GHz the magnitude deviation between any two weighting functions does not exceed 0.1dB. Conversely, the weighting functions for the second set spread over 3 dB for different distances.

In Fig. 5 we show the measured, normalized transfer factor for the set of 5 distances ranging from 100 to 140 cm. The reference distance that was set at 1 m. Fig. 6 shows the normalized, transfer factor measured for 10 equally spaced distances between 15 and 60 cm.

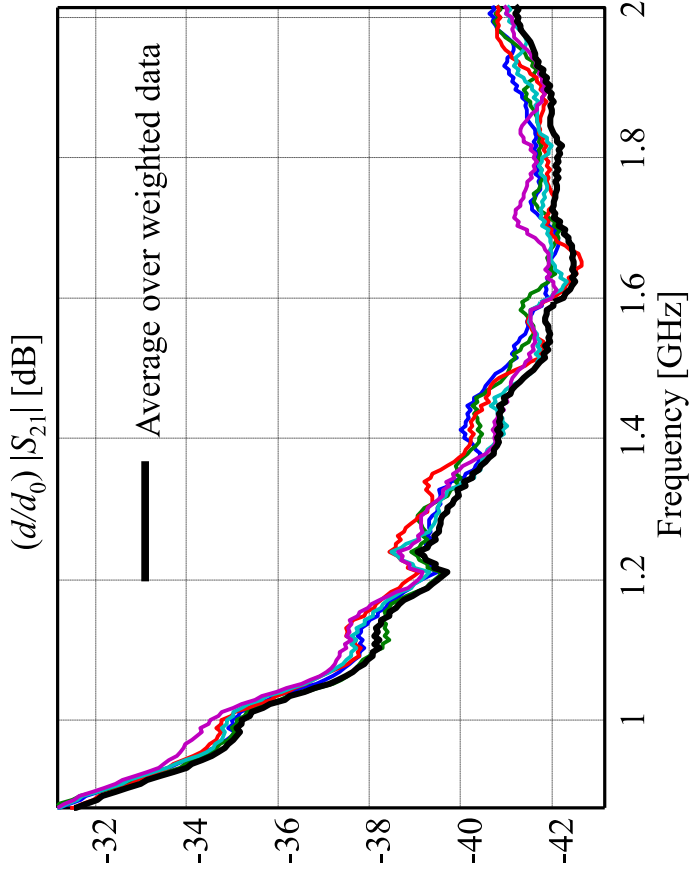


Fig.5 Normalized, transfer factors for $d=100$ to 140 cm

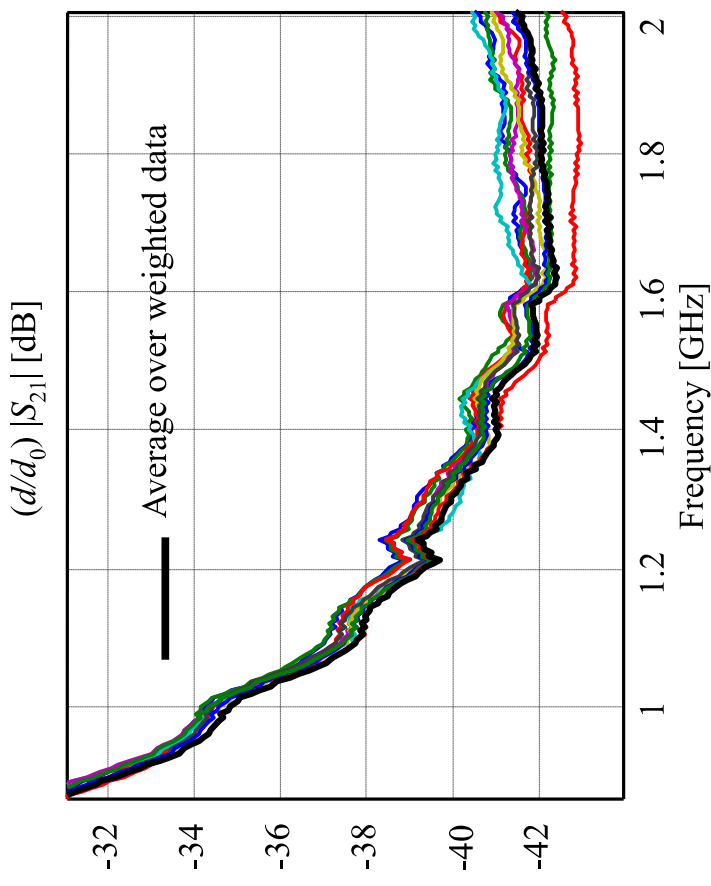


Fig.6 Normalized, transfer factors for $d=15$ to 60 cm

The gain figures resulting from the far-field set and from the near-field set are compared in Fig. 7. An unweighted average on the near-field data is also given, so as to observe the impact of the weighting on the accuracy. Table I gives the accuracy improvement defined as a difference between the relative error without weighting and the relative error with weighting; the far-field gain was set as a reference for computing both error figures.

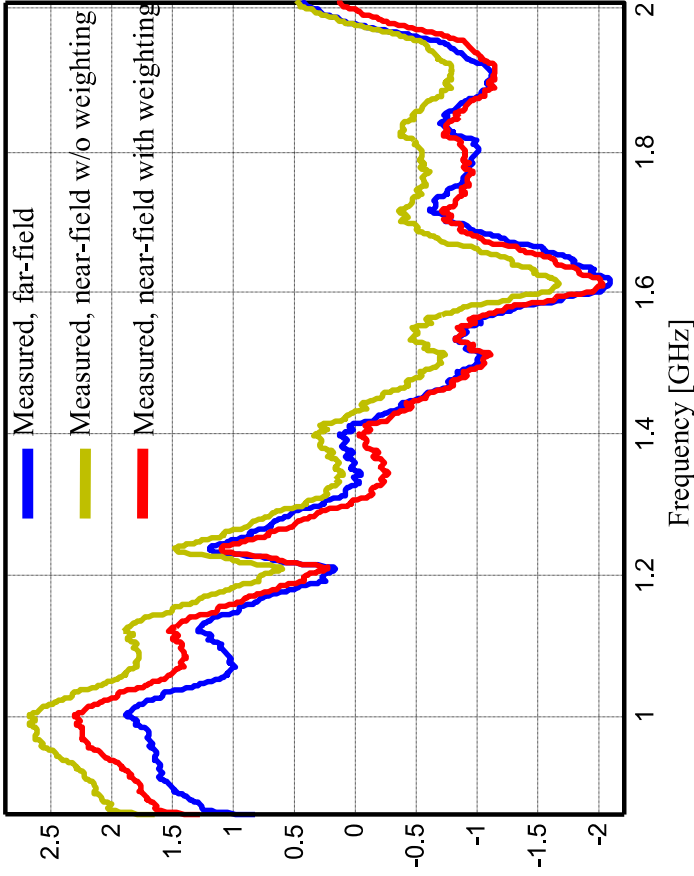


Fig. 7 Gain of the antenna under test

It comes out that weighting on the near-field data results in improving the accuracy of gain measurements by 0.5 dB for frequencies between 1 and 2 GHz.

Freq. [GHz]	Gain, far-field [dBi]	Gain, near-field [dBi]		Accuracy improvement [%]
		w/o weighting	w. weighting	
0.9	1.5	2.14	1.76	25.33
1	1.9	2.7	2.3	21.05
1.1	1.1	1.8	1.43	33.63
1.2	0.26	0.77	0.4	142.3
1.3	0.26	0.4	0.03	-34.6
1.4	0.113	0.326	-0.043	50.44
1.5	-1.01	-0.7	-1.068	24.95
1.6	-1.97	-1.54	-1.91	18.78
1.7	-0.8	-0.41	-0.775	45.62
1.8	-1	-0.55	-0.9	35
1.9	-1.093	-0.753	-1.1	30.46

Table I. Accuracy improvement on gain evaluation

CONCLUSION

- Our method can also be used for radiation pattern measurements with an appropriate two-axis rotator for the antenna under test.
- The distance averaging can be employed as a valid approach for gain measurements within a distance range entirely below the far-field limit.
- By applying the weighting functions that we introduced in this paper on measured, near-field data one can accurately assess the gain of an antenna, provided that the probe size is comparable to the wavelength.
- Our study was focused on antennas operating at around 1 GHz. The gain could be evaluated for frequencies of up to 2 GHz, based on measurements performed within a distance range of 15 to 60 cm, instead of placing the probe at above 100 cm as the fulfillment of the far-field constraints would require.
- Furthermore, if one would need to characterize antennas operating at around 100 MHz by applying our technique the measurements might be performed within a distance range of 1.5 m to 6 m.

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